

## FDP2532 / FDB2532

### N-Channel PowerTrench® MOSFET 150 V, 79 A, 16 mΩ

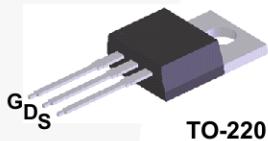
#### Features

- $R_{DS(on)} = 14 \text{ m}\Omega$  (Typ.) @  $V_{GS} = 10 \text{ V}$ ,  $I_D = 33 \text{ A}$
- $Q_{G(\text{tot})} = 82 \text{ nC}$  (Typ.) @  $V_{GS} = 10 \text{ V}$
- Low Miller Charge
- Low  $Q_{rr}$  Body Diode
- UIS Capability (Single Pulse and Repetitive Pulse)

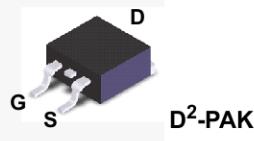
#### Applications

- Consumer Appliances
- Synchronous Rectification
- Battery Protection Circuit
- Motor drives and Uninterruptible Power Supplies
- Micro Solar Inverter

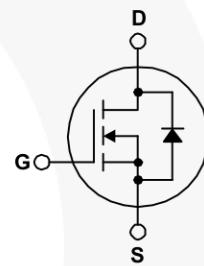
Formerly developmental type 82884



TO-220



D²-PAK



#### MOSFET Maximum Ratings $T_C = 25^\circ\text{C}$ unless otherwise noted

Symbol	Parameter	FDP2532 / FDB2532	Unit
$V_{DSS}$	Drain to Source Voltage	150	V
$V_{GS}$	Gate to Source Voltage	$\pm 20$	V
$I_D$	Drain Current Continuous ( $T_C = 25^\circ\text{C}$ , $V_{GS} = 10\text{V}$ )	79	A
	Continuous ( $T_C = 100^\circ\text{C}$ , $V_{GS} = 10\text{V}$ )	56	A
	Continuous ( $T_{amb} = 25^\circ\text{C}$ , $V_{GS} = 10\text{V}$ , $R_{\theta JA} = 43^\circ\text{C/W}$ )	8	A
	Pulsed	Figure 4	A
	Single Pulse Avalanche Energy (Note 1)	400	mJ
$P_D$	Power dissipation	310	W
	Derate above $25^\circ\text{C}$	2.07	W/ $^\circ\text{C}$
$T_J$ , $T_{STG}$	Operating and Storage Temperature	-55 to 175	$^\circ\text{C}$

#### Thermal Characteristics

$R_{\theta JC}$	Thermal Resistance Junction to Case, Max. TO-220, D²-PAK	0.61	$^\circ\text{C/W}$
$R_{\theta JA}$	Thermal Resistance Junction to Ambient, Max. TO-220, D²-PAK (Note 2)	62	$^\circ\text{C/W}$
$R_{\theta JA}$	Thermal Resistance Junction to Ambient D²-PAK, Max. 1in <sup>2</sup> copper pad area	43	$^\circ\text{C/W}$

## Package Marking and Ordering Information

Device Marking	Device	Package	Reel Size	Tape Width	Quantity
FDB2532	FDB2532	D <sup>2</sup> -PAK	330 mm	24 mm	800 units
FDP2532	FDP2532	TO-220	Tube	N/A	50 units

## Electrical Characteristics $T_C = 25^\circ\text{C}$ unless otherwise noted

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
<b>Off Characteristics</b>						
$B_{VDSS}$	Drain to Source Breakdown Voltage	$I_D = 250\mu\text{A}, V_{GS} = 0\text{V}$	150	-	-	V
$I_{DSS}$	Zero Gate Voltage Drain Current	$V_{DS} = 120\text{V}$ $V_{GS} = 0\text{V}$ $T_C = 150^\circ\text{C}$	-	-	1	$\mu\text{A}$
$I_{GSS}$	Gate to Source Leakage Current	$V_{GS} = \pm 20\text{V}$	-	-	$\pm 100$	nA
<b>On Characteristics</b>						
$V_{GS(TH)}$	Gate to Source Threshold Voltage	$V_{GS} = V_{DS}, I_D = 250\mu\text{A}$	2	-	4	V
$r_{DS(ON)}$	Drain to Source On Resistance	$I_D = 33\text{A}, V_{GS} = 10\text{V}$	-	0.014	0.016	$\Omega$
		$I_D = 16\text{A}, V_{GS} = 6\text{V},$	-	0.016	0.024	
		$I_D = 33\text{A}, V_{GS} = 10\text{V},$ $T_C = 175^\circ\text{C}$	-	0.040	0.048	

## Dynamic Characteristics

$C_{ISS}$	Input Capacitance	$V_{DS} = 25\text{V}, V_{GS} = 0\text{V},$ $f = 1\text{MHz}$	-	5870	-	pF
$C_{OSS}$	Output Capacitance		-	615	-	pF
$C_{RSS}$	Reverse Transfer Capacitance		-	135	-	pF
$Q_{g(TOT)}$	Total Gate Charge at 10V	$V_{GS} = 0\text{V to } 10\text{V}$ $V_{DD} = 75\text{V}$ $I_D = 33\text{A}$ $I_g = 1.0\text{mA}$	-	82	107	nC
$Q_{g(TH)}$	Threshold Gate Charge		-	11	14	nC
$Q_{gs}$	Gate to Source Gate Charge		-	23	-	nC
$Q_{gs2}$	Gate Charge Threshold to Plateau		-	13	-	nC
$Q_{gd}$	Gate to Drain "Miller" Charge		-	19	-	nC

## Resistive Switching Characteristics ( $V_{GS} = 10\text{V}$ )

$t_{ON}$	Turn-On Time	$V_{DD} = 75\text{V}, I_D = 33\text{A}$ $V_{GS} = 10\text{V}, R_{GS} = 3.6\Omega$	-	-	69	ns
$t_{d(ON)}$	Turn-On Delay Time		-	16	-	ns
$t_r$	Rise Time		-	30	-	ns
$t_{d(OFF)}$	Turn-Off Delay Time		-	39	-	ns
$t_f$	Fall Time		-	17	-	ns
$t_{OFF}$	Turn-Off Time		-	-	84	ns

## Drain-Source Diode Characteristics

$V_{SD}$	Source to Drain Diode Voltage	$I_{SD} = 33\text{A}$	-	-	1.25	V
		$I_{SD} = 16\text{A}$	-	-	1.0	V
$t_{rr}$	Reverse Recovery Time	$I_{SD} = 33\text{A}, dI_{SD}/dt = 100\text{A}/\mu\text{s}$	-	-	105	ns
$Q_{RR}$	Reverse Recovery Charge	$I_{SD} = 33\text{A}, dI_{SD}/dt = 100\text{A}/\mu\text{s}$	-	-	327	nC

### Notes:

- 1: Starting  $T_J = 25^\circ\text{C}$ ,  $L = 0.5\text{ mH}$ ,  $I_{AS} = 40\text{A}$ .
- 2: Pulse Width = 100s

**Typical Characteristics**  $T_C = 25^\circ\text{C}$  unless otherwise noted

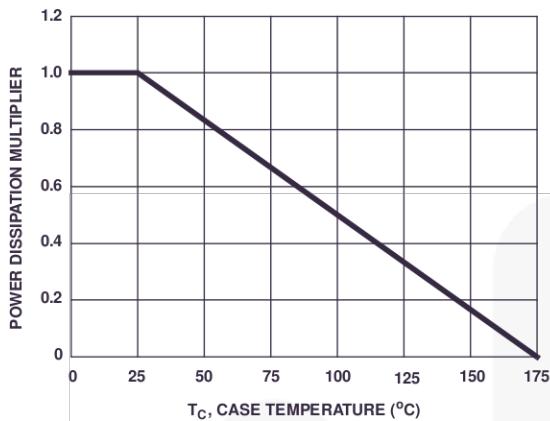


Figure 1. Normalized Power Dissipation vs Ambient Temperature

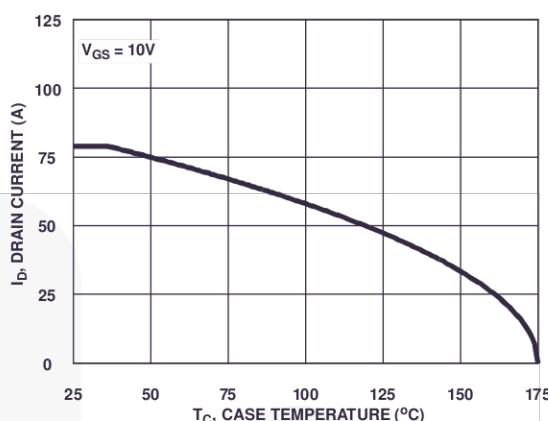


Figure 2. Maximum Continuous Drain Current vs Case Temperature

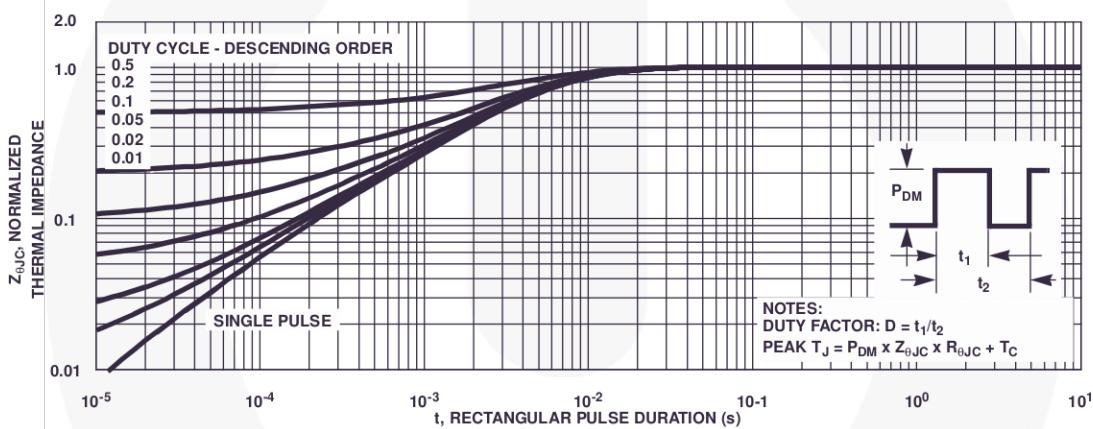


Figure 3. Normalized Maximum Transient Thermal Impedance

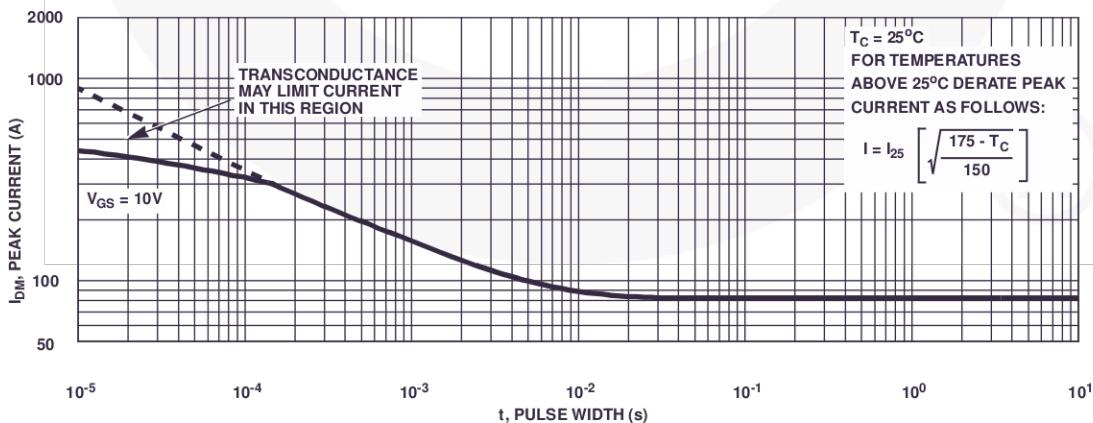
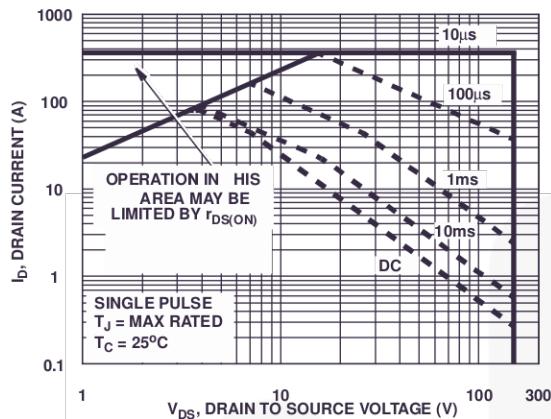
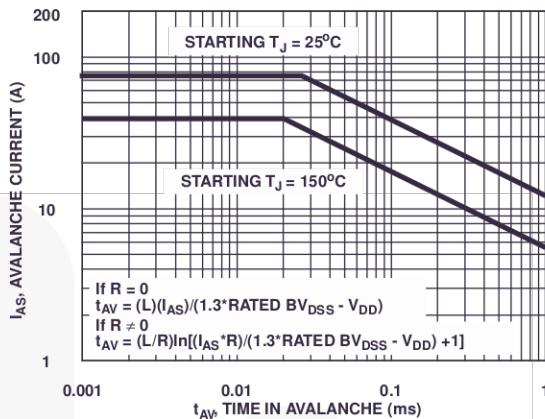


Figure 4. Peak Current Capability

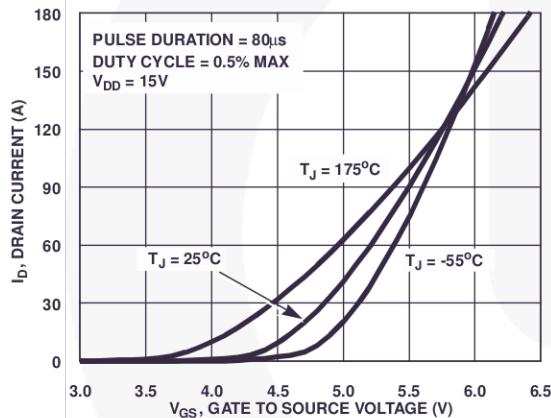
### Typical Characteristics $T_C = 25^\circ\text{C}$ unless otherwise noted



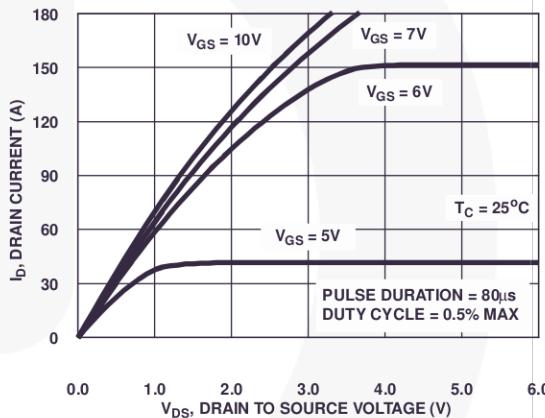
**Figure 5. Forward Bias Safe Operating Area**



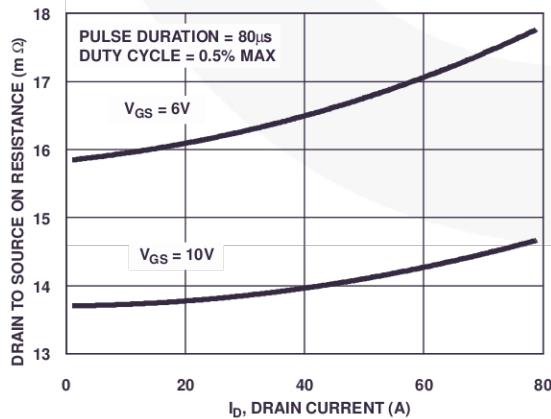
NOTE: Refer to Fairchild Application Notes AN7515 and AN7517  
**Figure 6. Unclamped Inductive Switching Capability**



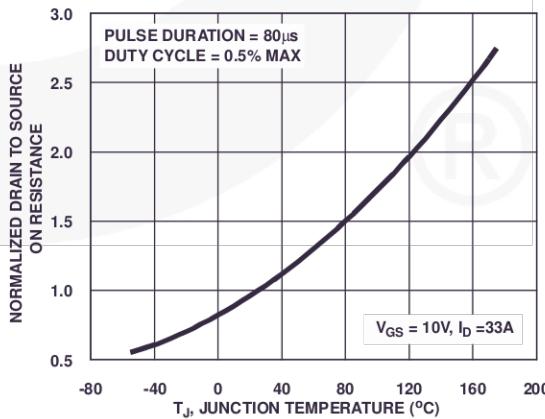
**Figure 7. Transfer Characteristics**



**Figure 8. Saturation Characteristics**



**Figure 9. Drain to Source On Resistance vs Drain Current**



**Figure 10. Normalized Drain to Source On Resistance vs Junction Temperature**

**Typical Characteristics**  $T_C = 25^\circ\text{C}$  unless otherwise noted

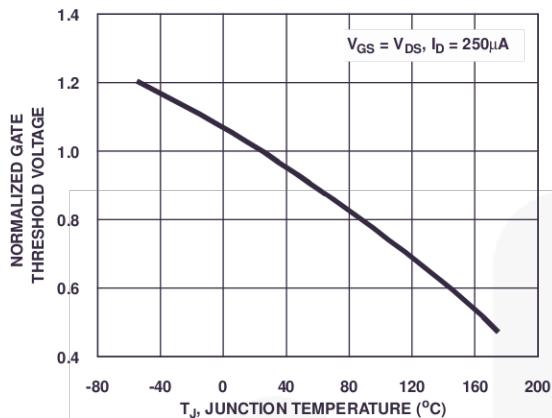


Figure 11. Normalized Gate Threshold Voltage vs Junction Temperature

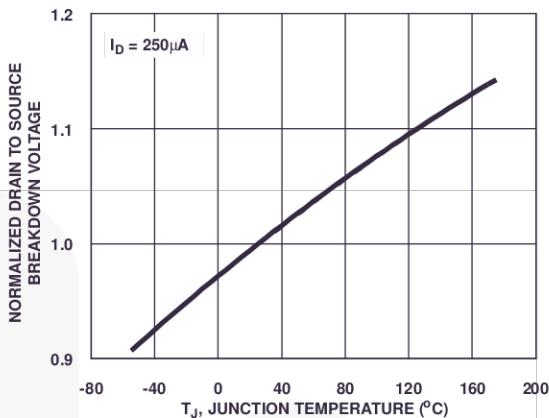


Figure 12. Normalized Drain to Source Breakdown Voltage vs Junction Temperature

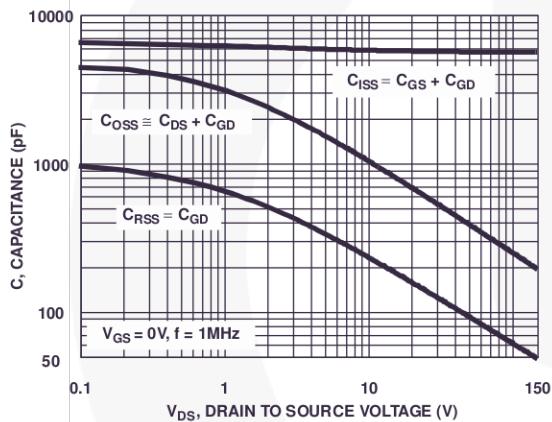


Figure 13. Capacitance vs Drain to Source Voltage

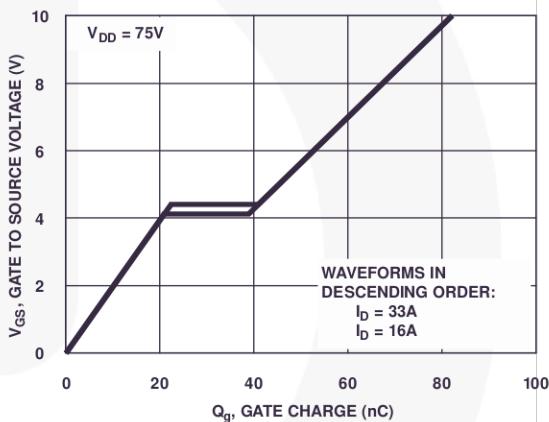


Figure 14. Gate Charge Waveforms for Constant Gate Currents

## Test Circuits and Waveforms

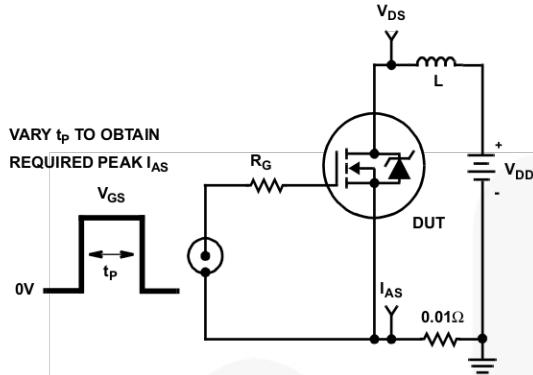


Figure 15. Unclamped Energy Test Circuit

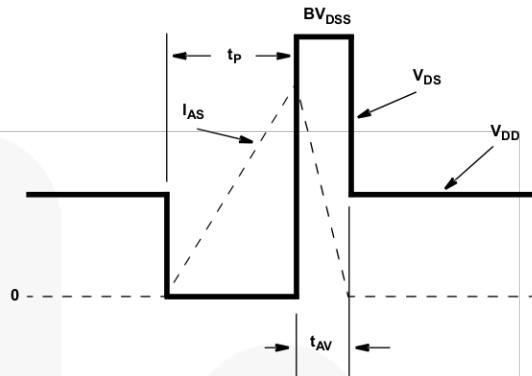


Figure 16. Unclamped Energy Waveforms

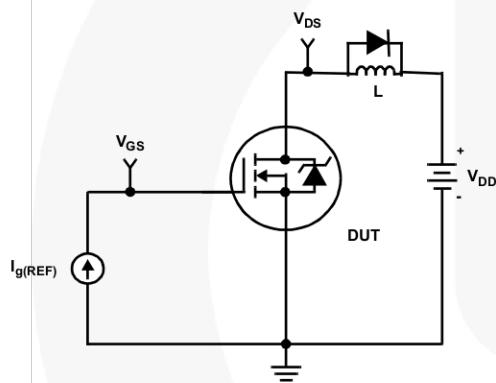


Figure 17. Gate Charge Test Circuit

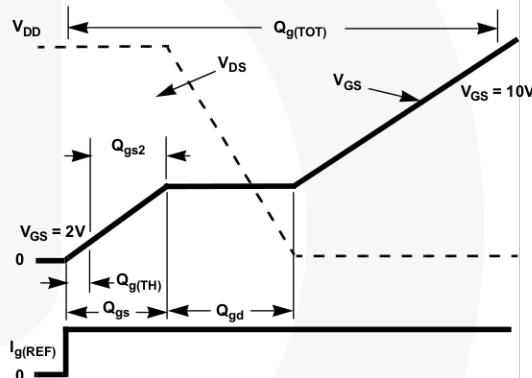


Figure 18. Gate Charge Waveforms

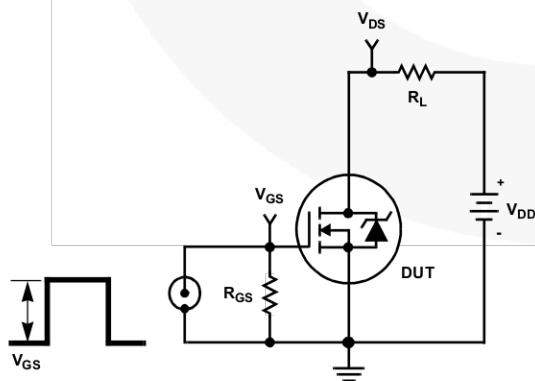


Figure 19. Switching Time Test Circuit

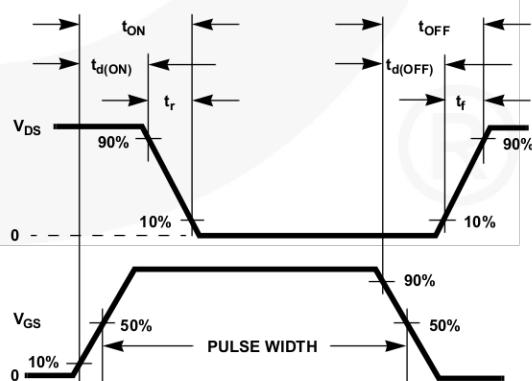


Figure 20. Switching Time Waveforms

## Thermal Resistance vs. Mounting Pad Area

The maximum rated junction temperature,  $T_{JM}$ , and the thermal resistance of the heat dissipating path determines the maximum allowable device power dissipation,  $P_{DM}$ , in an application. Therefore the application's ambient temperature,  $T_A$  ( $^{\circ}\text{C}$ ), and thermal resistance  $R_{\theta JA}$  ( $^{\circ}\text{C}/\text{W}$ ) must be reviewed to ensure that  $T_{JM}$  is never exceeded. Equation 1 mathematically represents the relationship and serves as the basis for establishing the rating of the part.

$$P_{DM} = \frac{(T_{JM} - T_A)}{R_{\theta JA}} \quad (\text{EQ. 1})$$

In using surface mount devices such as the TO-263 package, the environment in which it is applied will have a significant influence on the part's current and maximum power dissipation ratings. Precise determination of  $P_{DM}$  is complex and influenced by many factors:

1. Mounting pad area onto which the device is attached and whether there is copper on one side or both sides of the board.
2. The number of copper layers and the thickness of the board.
3. The use of external heat sinks.
4. The use of thermal vias.
5. Air flow and board orientation.
6. For non steady state applications, the pulse width, the duty cycle and the transient thermal response of the part, the board and the environment they are in.

Fairchild provides thermal information to assist the designer's preliminary application evaluation. Figure 21 defines the  $R_{\theta JA}$  for the device as a function of the top copper (component side) area. This is for a horizontally positioned FR-4 board with 1oz copper after 1000 seconds of steady state power with no air flow. This graph provides the necessary information for calculation of the steady state junction temperature or power dissipation. Pulse applications can be evaluated using the Fairchild device Spice thermal model or manually utilizing the normalized maximum transient thermal impedance curve.

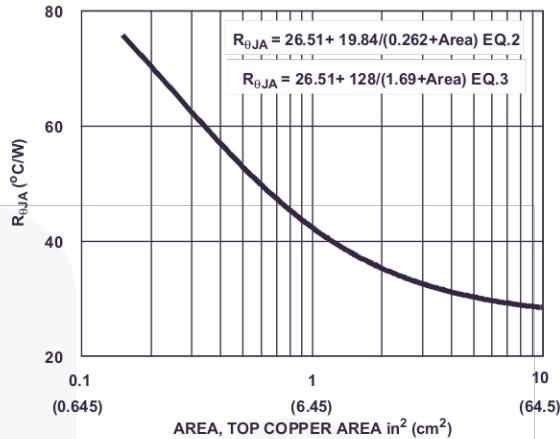
Thermal resistances corresponding to other copper areas can be obtained from Figure 21 or by calculation using Equation 2 or 3. Equation 2 is used for copper area defined in inches square and equation 3 is for area in centimeters square. The area, in square inches or square centimeters is the top copper area including the gate and source pads.

$$R_{\theta JA} = 26.51 + \frac{19.84}{(0.262 + \text{Area})} \quad (\text{EQ. 2})$$

Area in Inches Squared

$$R_{\theta JA} = 26.51 + \frac{128}{(1.69 + \text{Area})} \quad (\text{EQ. 3})$$

Area in Centimeters Squared



**Figure 21. Thermal Resistance vs Mounting Pad Area**

## **PSPICE Electrical Model**

.SUBCKT FDB2532 2 1 3 ; rev April 2002

CA 12 8 1.4e-9

CB 15 14 1.6e-9

GIN 6 8 5.61e-9

Dbody 7 5 DbodyMOD  
Dbreak 5 11 DbreakMOD  
Dplcap 10 5 DplcapMOD

Ebreak	11	7	17	18	159
Eds	14	8	5	8	1
Egs	13	8	6	8	1
Esg	6	10	6	8	1
Evthres	6	21	19	8	1
Evtemp	20	6	18	22	1

It 8 17 1

Lgate 1 9 9.56e-9  
Ldrain 2 5 1.0e-9  
Lsource 3 7 7.71e-9

RLgate 1 9 95.6  
RLdrain 2 5 10  
RLsource 3 7 77.1

Mmed 16 6 8 8 MmedMOD  
Mstro 16 6 8 8 MstroMOD  
Mweak 16 21 8 8 MweakMOD

```
Rbreak 17 18 RbreakMOD 1
Rdrain 50 16 RdrainMOD 9.6e-3
Rgate 9 20 1.01
RSLC1 5 51 RSLCMOD 1.0e-6
RSLC2 5 50 1.0e3
Rsource 8 7 RsourceMOD 3.0e-3
Rvthres 22 8 RvthresMOD 1
Rvttemp 18 19 RvttempMOD 1
S1a 6 12 13 8 S1AMOD
S1b 13 12 13 8 S1BMOD
S2a 6 15 14 13 S2AMOD
S2b 10 15 11 12 S2BMOD
```

What 22 19 DC 1

ESI C 51.50 VALUE = {((V(5.51)/ABS(V(5.51))))\*(PWB(V(5.51)/(1e-6\*190.3)))}

.MODEL DbodyMOD D (IS=6.0E-11 N=1.09 RS=2.3e-3 TRS1=3.0e-3 TRS2=1.0e-6  
+ C1Q=3.9e-9 M=0.65 TT=4.8e-8 XT1=4.2)

```
+ CJO=3.9e-9 M=0.65 T1=4.8e-6 T1I=4.2)
.MODEL DbreakMOD D (RS=0.17 TRS1=3.0e-3 TRS2=-8.9e-6)
.MODEL DplcapMOD D (CJO=1.0e-9 IS=1.0e-30 N=10 M=0.6)
```

MODEL MmedMOD NMOS (VTO=-3.55 KP=10 IS=1e-30 N=10 TOX=1 L=1u W=1u RG=1.01)

```

.MODEL MfmedMOD NMOS (VTO=3.55 KP=10 IS=1e-30 N=10 TOX=1 L=10 W=10)
.MODEL MstromMOD NMOS (VTO=-4.2 KP=145 IS=1e-30 N=10 TOX=1 L=10 W=10)

```

```
.MODEL MweakMOD NMOS (VTO=4.2 KP=14 IS=1e-30 N=10 TOX=1 L=10 W=10)
.MODEL MweakMOD NMOS (VTO=-2.9 KP=0.05 IS=1e-30 N=10 TOX=1 L=1 W=1 RG=10.1 RS=0.1)
```

```

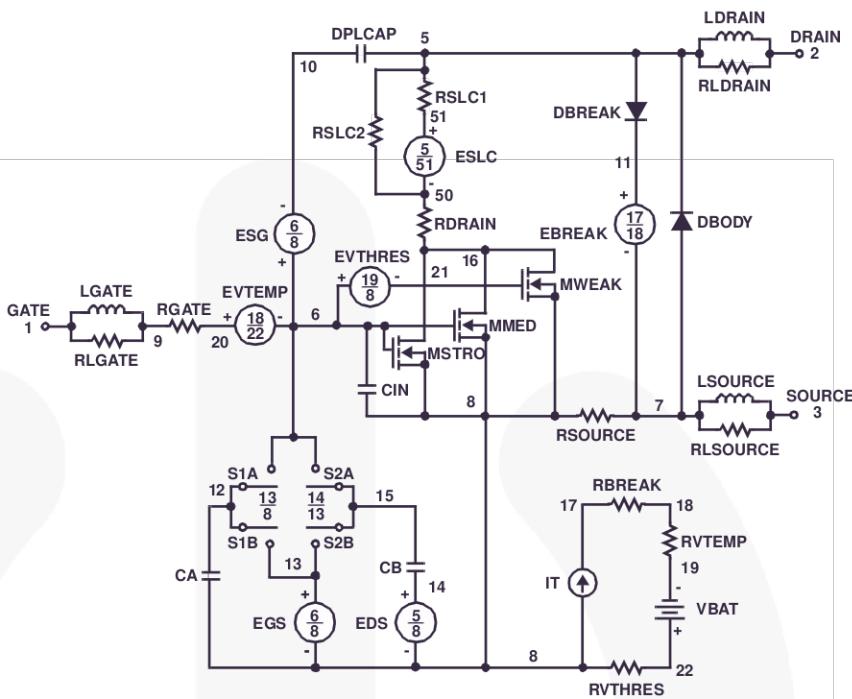
.MODEL RbreakMOD RES (TC1=1.1e-3 TC2=-9.0e-7)
.MODEL RdrainMOD RES (TC1=9.0e-3 TC2=3.5e-5)
.MODEL RSCLMOD RES (TC1=3.4e-3 TC2=1.5e-6)
.MODEL RsourceMOD RES (TC1=4.0e-3 TC2=1.0e-6)
.MODEL RVthresMOD RES (TC1=-4.1e-3 TC2=-1.4e-5)
.MODEL RymemMOD RES (TC1=-4.0e-3 TC2=3.5e-6)

```

```
.MODEL S1AMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-6.0 VOFF=-4.0)
.MODEL S1BMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-4.0 VOFF=-6.0)
.MODEL S2AMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-1.4 VOFF=1.0)
.MODEL S2RMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=1.0 VOFF=-1.4)
```

ENDS

.ENDS  
Note: For further discussion of the PSPICE model, consult **A New PSPICE Sub-Circuit for the Power MOSFET Featuring Global Temperature Options**; IEEE Power Electronics Specialist Conference Records, 1991, written by William J. Hepp and C. Frank Wheatley.



## SABER Electrical Model

```

REV April 2002
ttemplate FDB2532 n2,n1,n3
electrical n2,n1,n3
{
var i iscl
dp..model dbodymod = (isl=6.0e-11,nl=1.09,rs=2.3e-3,trs1=3.0e-3,trs2=1.0e-6,cjo=3.9e-9,m=0.65,tt=4.8e-8,xti=4.2)
dp..model dbreakmod = (rs=0.17,trs1=3.0e-3,trs2=-8.9e-6)
dp..model dplcapmod = (cjo=1.0e-9,isl=10.0e-30,nl=10,m=0.6)
m..model mmedmod = (type=_n,vto=3.55,kp=10,is=1e-30,tox=1)
m..model mstrongmod = (type=_n,vto=4.2,kp=145,is=1e-30,tox=1)
m..model mweakmod = (type=_n,vto=2.9,kp=0.05,is=1e-30,tox=1,rs=0.1)
sw_vcsp..model s1amod = (ron=1e-5,roff=0.1,von=-6.0,voff=-4.0)
sw_vcsp..model s1bmod = (ron=1e-5,roff=0.1,von=-4.0,voff=-6.0)
sw_vcsp..model s2amod = (ron=1e-5,roff=0.1,von=-1.4,voff=1.0)
sw_vcsp..model s2bmod = (ron=1e-5,roff=0.1,von=1.0,voff=-1.4)
c.ca n12 n8 = 1.4e-9
c.cb n15 n14 = 1.6e-9
c.cin n6 n8 = 5.61e-9

dp.dbody n7 n5 = model=dbodymod
dp.dbreak n5 n11 = model=dbreakmod
dp.dplcap n10 n5 = model=dplcapmod

spe.ebreak n11 n7 n17 n18 = 159
GATE 1 o LGATE
RGATE 9 o RLGATE
ESG 6 8
EVTTEMP 18 22
EVTHRES 19 8
MSTRO 16 21
MMED 17 21
MWEAK 18 21
EBREAK 17 18
LSOURCE 7 o SOURCE 3
RLSOURCE 7 o RLSOURCE 3
IT 17
RBREAK 17 18
RVTEMP 19
RVTHRES 22
VBAT 22

i.it n8 n17 = 1
I.igate n1 n9 = 9.56e-9
I.ldrain n2 n5 = 1.0e-9
I.lssource n3 n7 = 7.71e-9

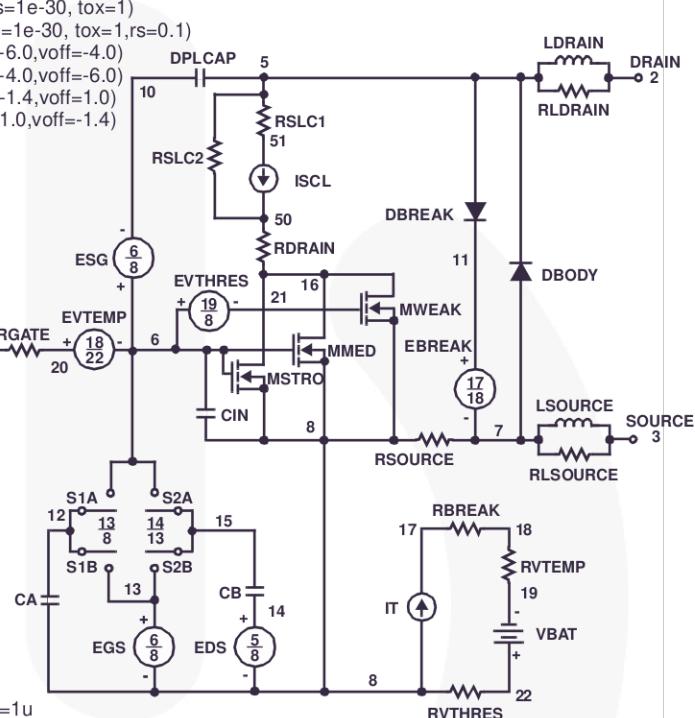
res.rigate n1 n9 = 95.6
res.rldrain n2 n5 = 10
res.rlsource n3 n7 = 77.1

m.mmed n16 n6 n8 n8 = model=mmedmod, l=1u, w=1u
m.mstrong n16 n6 n8 n8 = model=mstrongmod, l=1u, w=1u
m.mweak n16 n21 n8 n8 = model=mweakmod, l=1u, w=1u

res.rbreak n17 n18 = 1, tc1=1.1e-3,tc2=-9.0e-7
res.rdrain n50 n16 = 9.6e-3, tc1=9.0e-3,tc2=3.5e-5
res.rgate n9 n20 = 1.01
res.rsclc1 n5 n51 = 1.0e-6, tc1=3.4e-3,tc2=1.5e-6
res.rsclc2 n5 n50 = 1.0e3
res.rsource n8 n7 = 3.0e-3, tc1=4.0e-3,tc2=1.0e-6
res.rvthres n22 n8 = 1, tc1=-4.1e-3,tc2=-1.4e-5
res.rvtemp n18 n19 = 1, tc1=-4.0e-3,tc2=3.5e-6
sw_vcsp.s1a n6 n12 n13 n8 = model=s1amod
sw_vcsp.s1b n13 n12 n13 n8 = model=s1bmod
sw_vcsp.s2a n6 n15 n14 n13 = model=s2amod
sw_vcsp.s2b n13 n15 n14 n13 = model=s2bmod

v.vbat n22 n19 = dc=1
equations {
i (n51->n50) +=iscl
iscl: v(n51,n50) = ((v(n5,n51)/(1e-9+abs(v(n5,n51))))*((abs(v(n5,n51)*1e6/190))** 3))
}
}

```



## SPICE Thermal Model

REV 26 February 2002

FDB2532

```
CTHERM1 TH 6 7.5e-3
CTHERM2 6 5 8.0e-3
CTHERM3 5 4 9.0e-3
CTHERM4 4 3 2.4e-2
CTHERM5 3 2 3.4e-2
CTHERM6 2 TL 6.5e-2
```

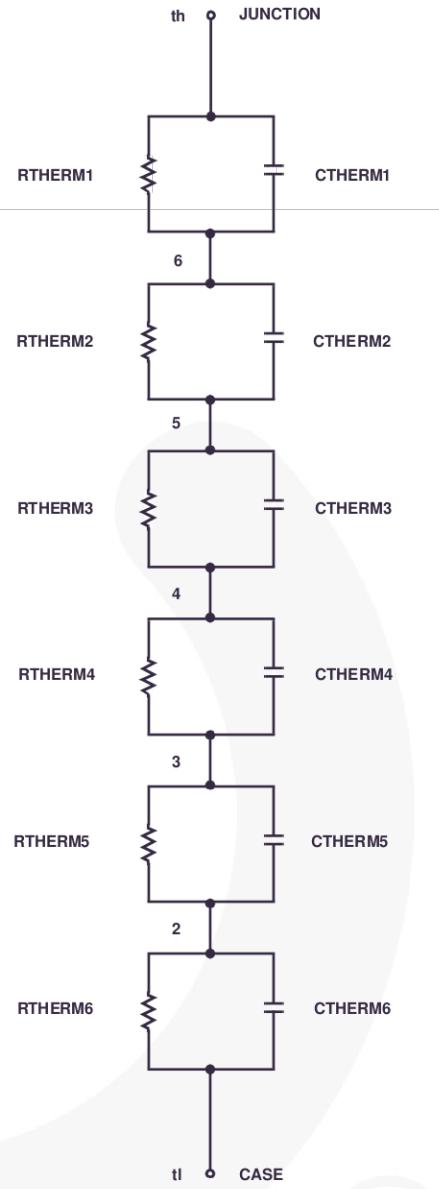
```
RTERM1 TH 6 3.1e-4
RTERM2 6 5 2.5e-3
RTERM3 5 4 2.0e-2
RTERM4 4 3 8.0e-2
RTERM5 3 2 1.2e-1
RTERM6 2 TL 1.3e-1
```

## SABER Thermal Model

SABER thermal model FDB2532  
template thermal\_model th tl

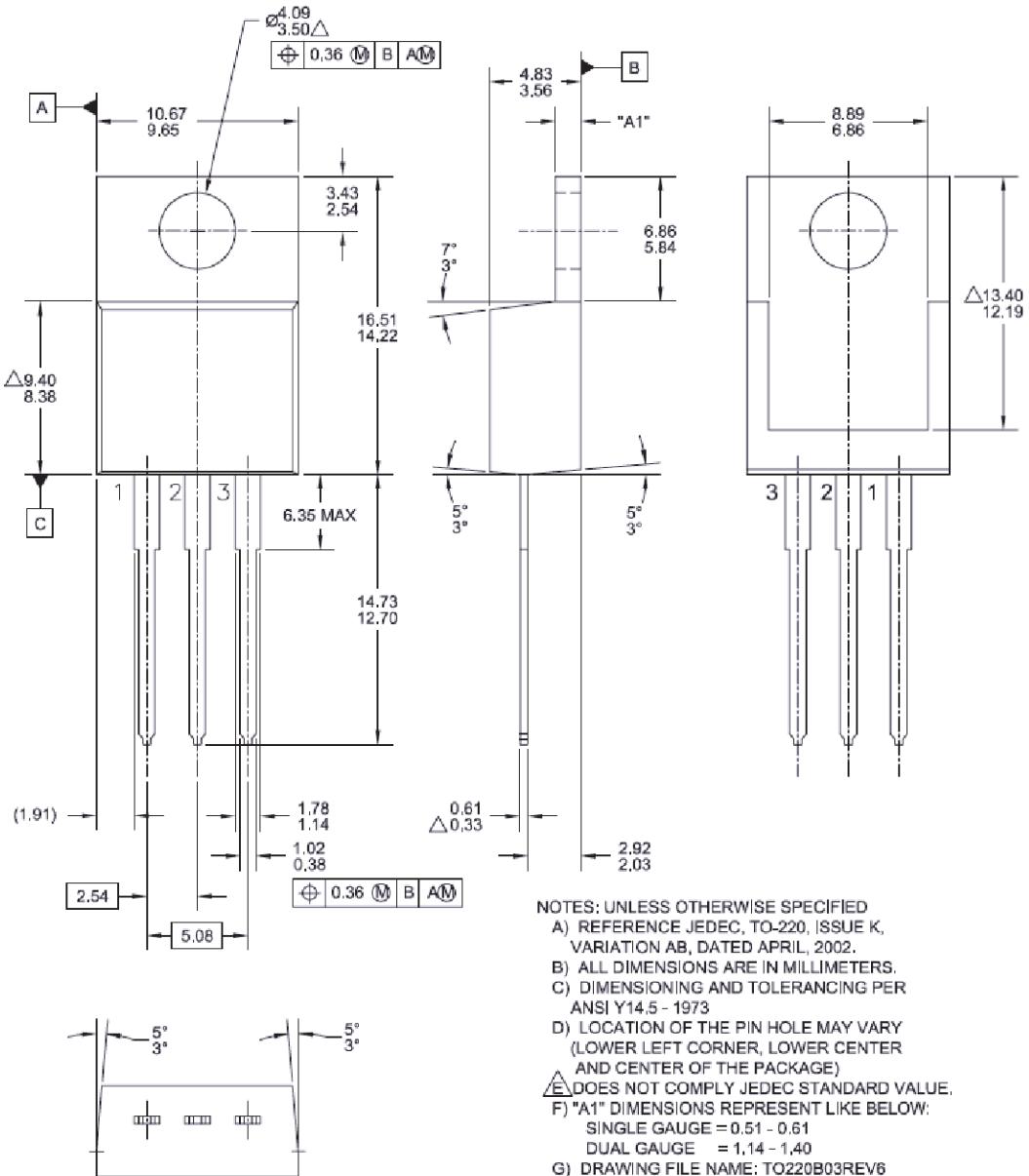
```
thermal_c th, tl
{
    ctherm.ctherm1 th 6 =7.5e-3
    ctherm.ctherm2 6 5 =8.0e-3
    ctherm.ctherm3 5 4 =9.0e-3
    ctherm.ctherm4 4 3 =2.4e-2
    ctherm.ctherm5 3 2 =3.4e-2
    ctherm.ctherm6 2 tl =6.5e-2

    rrtherm.rtherm1 th 6 =3.1e-4
    rtherm.rtherm2 6 5 =2.5e-3
    rtherm.rtherm3 5 4 =2.0e-2
    rtherm.rtherm4 4 3 =8.0e-2
    rtherm.rtherm5 3 2 =1.2e-1
    rtherm.rtherm6 2 tl =1.3e-1
}
```



## Mechanical Dimensions

### TO-220 3L



**Figure 22. TO-220, Molded, 3Lead, Jedec Variation AB**

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Dimension in Millimeters

## Mechanical Dimensions

### TO-263 2L (D<sup>2</sup>PAK)

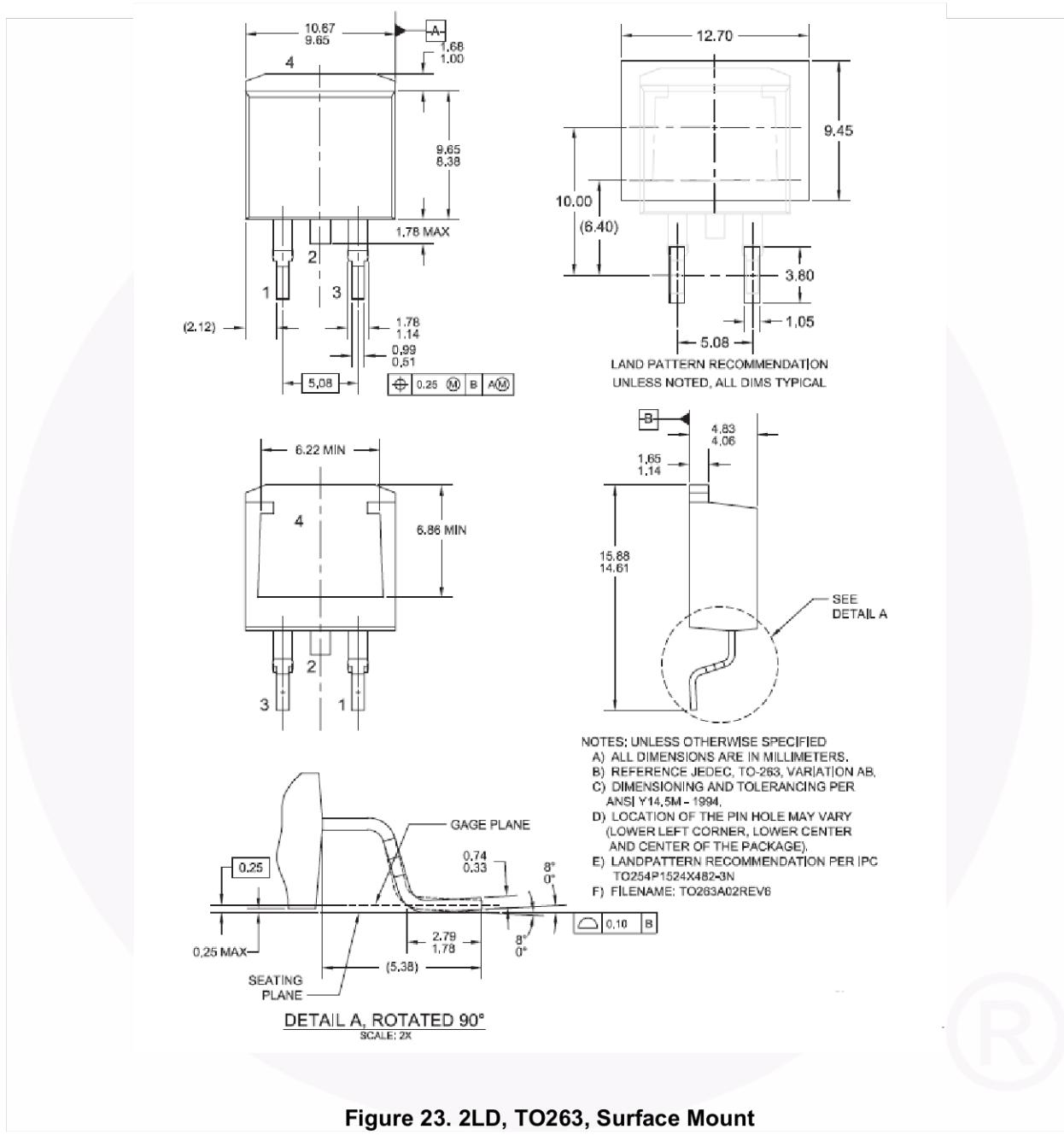


Figure 23. 2LD, TO263, Surface Mount

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Dimension in Millimeters



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SuperFET®  
SuperSOT™-3  
SuperSOT™-6  
SuperSOT™-8  
SupreMOS®  
SyncFET™

Sync-Lock™  
 SYSTEM®\*  
GENERAL  
TinyBoost®  
TinyBuck®  
TinyCalc™  
TinyLogic®  
TINYOPTO™  
TinyPower™  
TinyPWM™  
TinyWire™  
TransiC™  
TriFault Detect™  
TRUECURRENT®\*  
μSerDes™  
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